

DNA 4615T

SYNTHESIS OF A PLUME SIMULATOR FOR THE MX MISSILE

Dikewood Industries, Inc. 1100 Glendon Avenue Los Angeles, California 90024

31 May 1978

Topical Report for Period 7 November 1977-31 May 1978

Subcontract CONTRACT No. DNA 001-77-C-0184

USE SAI, Berkeley, CA 409694

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	REPORT DOCUMENTATION PAGE		
1. REPORT NUMBER	2. GOVT ACCESSION NO.	3. RECIPIENT'S CATALOG NUMBER	
DNA 4615T			
1. TITLE (and Subtitle)		5. TYPE OF REPORT & PERIOD COVERED	
		Topical Report for Period	
SYNTHESIS OF A PLUME SIMULATOR FOR THE MX MISSILE		7 Nov 77—31 May 78	
		6. PERFORMING ORG. REPORT NUMBER	
7. AUTHOR(s)		8. CONTRACT OR GRANT NUMBER(s)	
John Lam		DNA 001-77-C-0184	
9. PERFORMING ORGANIZATION NAME AND ADDRE	ESS	10. PROGRAM ELEMENT, PROJECT, TASK AREA & WORK UNIT NUMBERS	
Dikewood Industries, Inc.			
1100 Glendon Avenue		Subtask R99QAXEB088-11	
Los Angeles, California 90024			
1. CONTROLLING OFFICE NAME AND ADDRESS		12. REPORT DATE	
Director		31 May 1978	
Defense Nuclear Agency		13. NUMBER OF PAGES	
Washington, D.C. 20305		15. SECURITY CLASS (of this report)	
14. MONITORING AGENCY NAME & ADDRESS(II dille	erent from Controlling Office)		
		UNCLASSIFIED	
		15a. DECLASSIFICATION/DOWNGRADING SCHEDULE	
16. DISTRIBUTION STATEMENT (of this Report)			
Approved for public release; di	istribution unlimit	ced.	
17. DISTRIBUTION STATEMENT (of the abstract enter	red in Block 20, if different from	m Report)	

18. SUPPLEMENTARY NOTES

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19. KEY WORDS (Continue on reverse side if necessary and identify by block number)

Electromagnetic pulse

Missile Plume Modeling Missile Plume Simulation

EMP Coupling

EMP

MX Missile

EMP Interaction

EMP Simulation

20. ABSTRACT (Continue on reverse side if necessary and identify by block number)

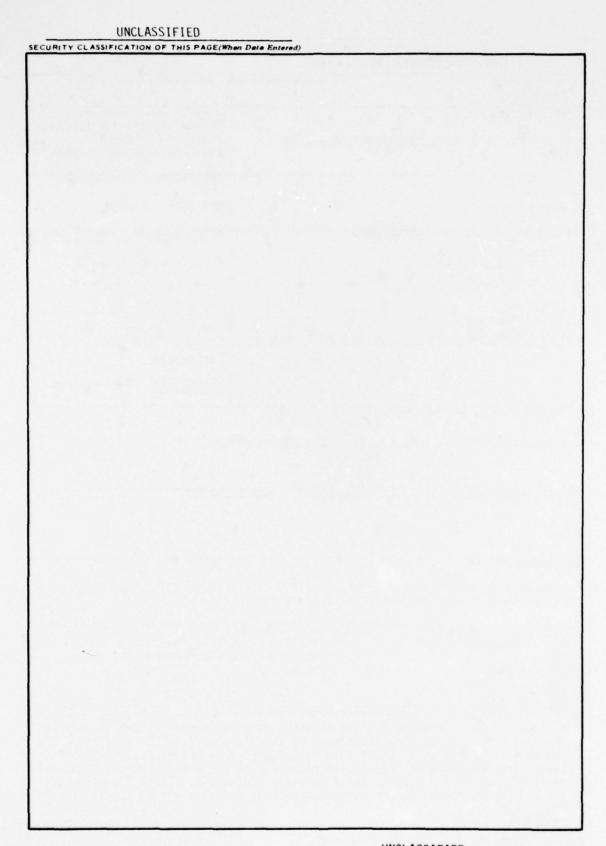
> A simulator for the exhaust plume of an MX missile is synthesized for use in the in-flight EMP hardness assessment of the missile in an EMP simulator. The synthesis is based on a set of results obtained in a numerical analysis of the interaction between the missile plume and an EMP which uses the Titan-III missile plume data scaled to the MX missile dimensions. It is found that, for this case, the plume simulator can be made entirely out of passive lumped LRC elements.

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I. INTRODUCTION

A ballistic missile in free flight emits continuously an enormous volume of exhaust gases from its engine nozzle. This exhaust is trailed behind the missile in the form of a gaseous plume. Within the earth's atmosphere the visible part of the plume has the appearance of a long, luminous column extending beyond the nozzle to a length of several times that of the missile.

The plume contains a high concentration of ions and electrons. The presence of these charged particles in the missile's immediate environment can effectively alter the electromagnetic characteristics of the missile. For example, if the plumed missile were to be struck by the electromagnetic pulse (EMP) of a nuclear explosion, as illustrated in Fig. 1, the electric current and charge induced by the EMP on the missile skin would assume different values than if the plume were absent.

The interaction of the MX missile in free flight with an EMP has been analyzed with a theoretical model [1]. The model employed the Titan-III missile plume data at an altitude of 20 kilometers and scaled to the MX missile dimensions. The EMP waveform was assumed of the double-exponential type. The total induced current flowing up from the plume onto the missile at the nozzle during the EMP encounter was calculated. Fig. 2 shows one set of the results of the calculation. The nozzle current $I_{\rm noz}$ in the frequency domain is here plotted versus the frequency f for broadside EMP incidence. The incident electric field is taken to be polarized along the length of the missile, and at a strength of 1 volt per meter.

When the missile is tested for its in-flight EMP hardness in an EMP simulator such as ARES, it is necessary to make up for the absence of the plume at testing with a plume simulator. One method of plume simulation is illustrated in Fig. 3. The figure shows a missile standing upright in an EMP simulator and struck by the simulator field. If the simulator field is a good simulation of the EMP, then the state of affairs above the nozzle duplicates correctly that in the in-flight EMP encounter depicted in Fig. 1.

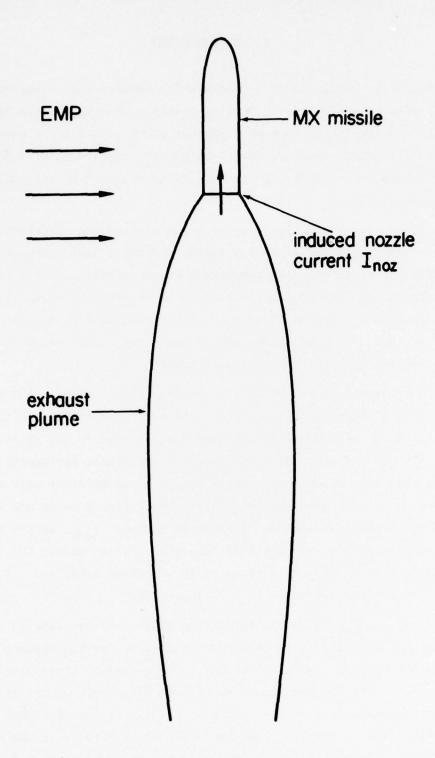


Fig. 1. Interaction of an EMP with an MX missile and its exhaust plume.

The plume picks up an electric current and injects it onto the missile skin at the nozzle.

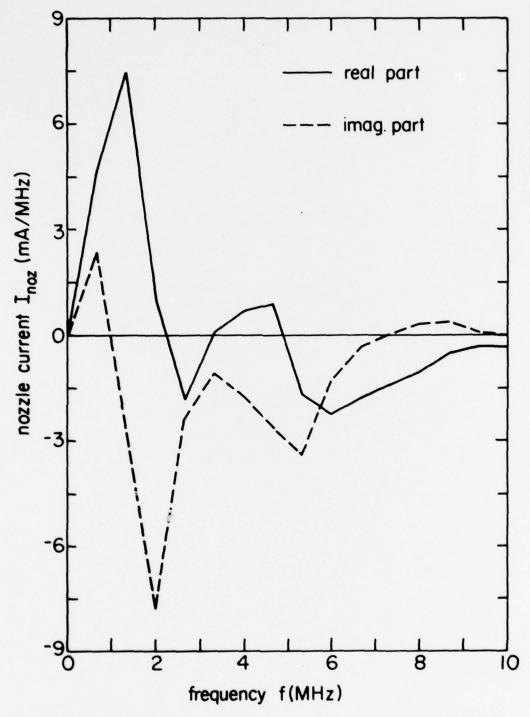


Fig. 2. Nozzle current on the MX missile in the frequency domain for broadside EMP incidence, as calculated in Ref. [1]. The incident electric field is directed along the length of the missile and of intensity 1 volt per meter.

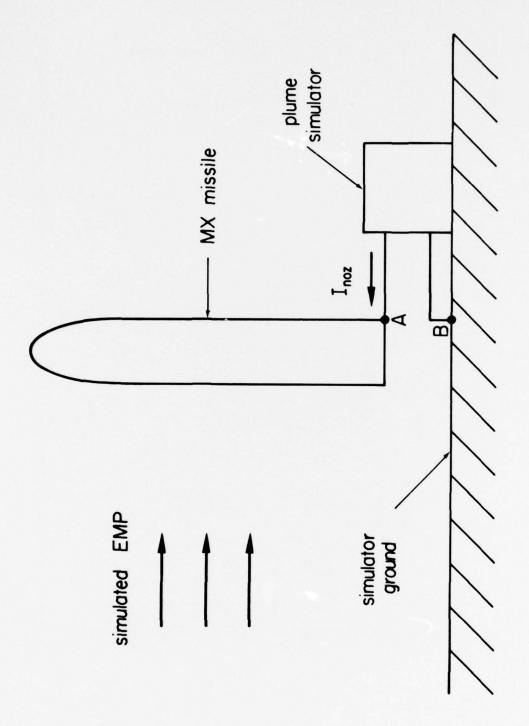


Fig. 3. Geometry of an MX missile under test in an EMP simulator.

However, the same cannot be said of the situation below the nozzle. Instead of the plume, one finds here the missile's image with respect to the simulator ground. Therefore one attempts to compensate for this difference by connecting a plume simulator between the missile nozzle and ground, that is, across the terminals A and B in Fig. 3. This plume simulator is to be so designed as to deliver, in conjunction with the simulator field, a net current to the terminal A equal to the nozzle current I induced by the EMP.

The following sections present the parametrization and synthesis of the plume simulator.

II. PARAMETRIZATION OF THE PLUME SIMULATOR

The interaction of the EMP simulator field with the missile sets up electric currents on the missile skin and the simulator ground. As seen at the two terminals A and B in Fig. 3, this electromagnetic excitation of the missile can be described by a Thévenin equivalent circuit shown in Fig. 4. The circuit consists of an open-circuit voltage V_{oc} and an input impedance Z_{in} connected in series.

Under the assumption of a perfectly-conducting simulator ground, the input admittance Y_{in} (= $1/Z_{in}$) across the terminals A and B has been calculated for the MX missile [1]. The results are plotted in Fig. 5. If the terminals A and B are short-circuited, a short-circuit current I_{sc} will flow from terminal B to terminal A . It is given by

$$I_{sc} = V_{oc}/Z_{in} = Y_{in}V_{oc}$$
 (1)

This current has been calculated for the case of a broadside incident simulated EMP whose electric field is vertically polarized and at a strength of 1 volt per meter [1]. The results are plotted in Fig. 6.

In a similar fashion the plume simulator of Fig. 3 can be represented by a Thévenin equivalent circuit between the terminals A and B, consisting of an open-circuit voltage V_s and an input impedance Z_s . When the plume simulator is hocked up to the missile during testing, the entire test system consisting of the EMP simulator, the missile and the plume simulator is described by the closed circuit shown in Fig. 7. The plume simulator parameters V_s and Z_s must be so chosen that the current flowing in the circuit -- and hence through the terminal A onto the missile -- is equal to the nozzle current I_{noz} given in Fig. 2.

The circuit in Fig. 7 yields the relation

$$(Z_{in} + Z_s)I_{noz} = V_{oc} + V_s$$
 (2)

Solving for Z_s , one has

$$Z_{s} = Z_{1} + Z_{2}$$
 (3)

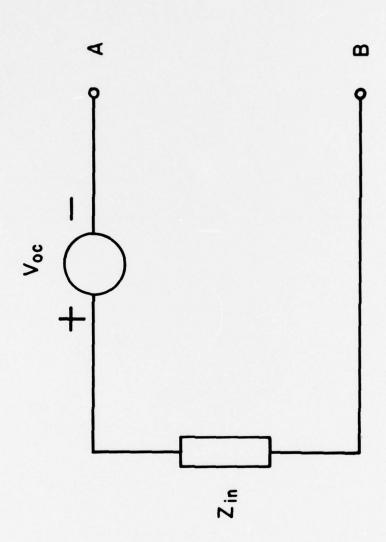


Fig. 4. Thevenin equivalent circuit representing a missile excited by an EMP.

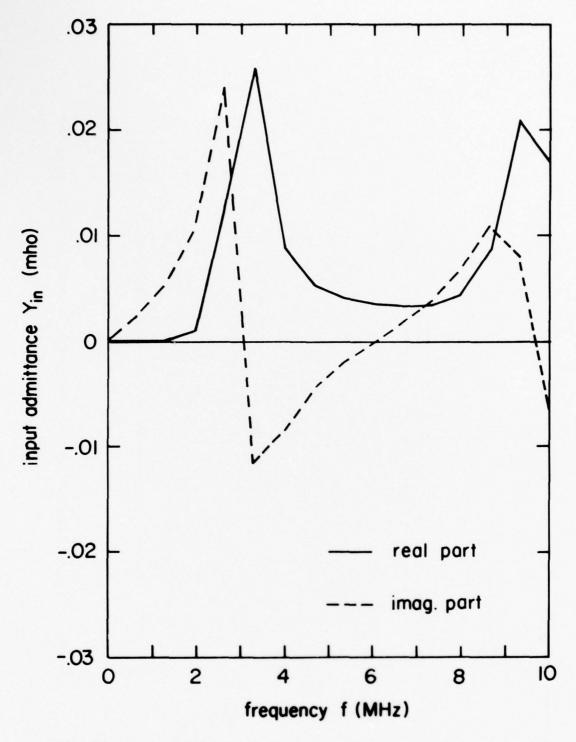


Fig. 5. Input admittance of an upright MX missile relative to a perfectly-conducting ground, as calculated in Ref. [1].

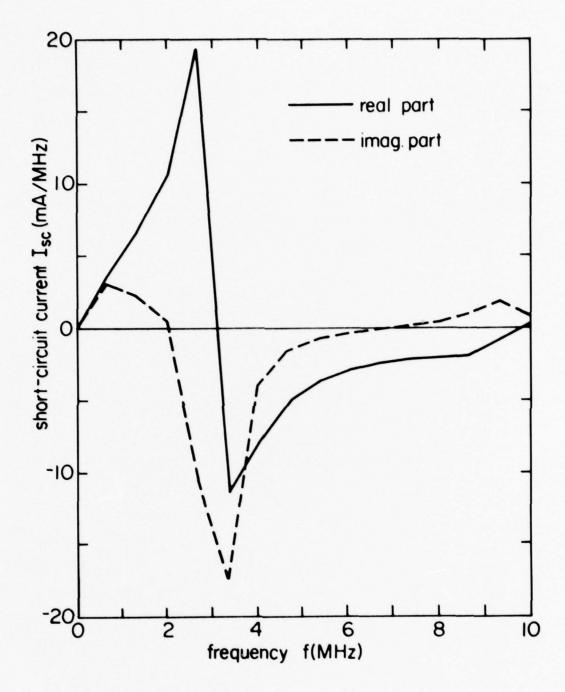
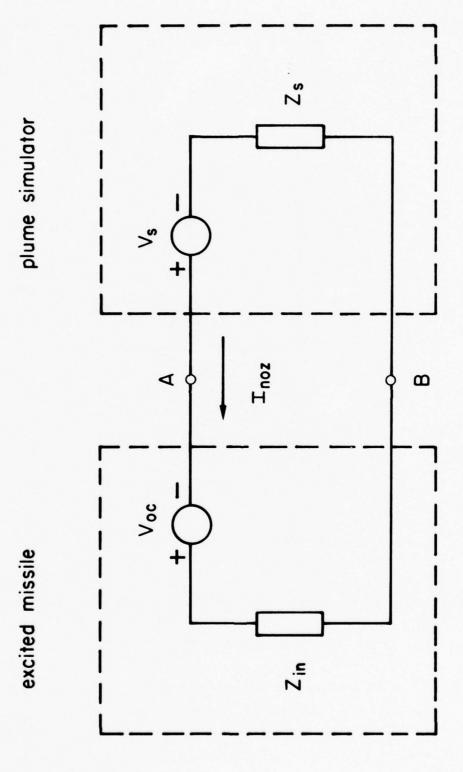


Fig. 6. Short-circuit current flowing from ground to the MX missile nozzle for broadside simulator EMP incidence, as calculated in Ref. [1]. The simulator electric field is vertically polarized and of intensity 1 volt per meter.



Equivalent circuit for a missile under test in an EMP simulator and connected to a plume simulator. Fig. 7.

where

$$Z_{1} = \frac{v_{oc}}{I_{noz}} - Z_{in} = \left(\frac{I_{sc}}{I_{noz}} - 1\right) Z_{in}$$
 (4)

and

$$Z_2 = \frac{V_s}{I_{noz}} \tag{5}$$

 $\mathbf{Z}_{\mathbf{S}}$ is a passive network element. Its real part must satisfy the requirement of being nonnegative.

Consider first Z_1 , or rather its reciprocal the admittance Y_1 :

$$Y_1 = \frac{1}{Z_1} = \frac{Y_{in}}{I_{sc}/I_{noz} - 1}$$
 (6)

From the values of I_{noz} , Y_{in} and I_{sc} in Figs. 2,5 and 6, one can evaluate Y_1 . The results are shown in Fig. 8. The most noticeable feature of Y_1 is that its real part is practically nonnegative up to nearby 10 MHz, with the negligible exception of a very small dip below 0 between 5 and 6 MHz. How Y_1 behaves above 10 MHz is here immaterial, since the currents I_{noz} and I_{sc} do not have significant components in this high-frequency region.

The fact that the real part of Y_1 is positive in the frequency range of interest has certain far-reaching implications. First, it implies that Y_1 itself can be realized with passive network elements alone. More important, it implies that the active element V_s in the plume simulator equivalent circuit is really redundant. The plume simulator can be made simply out of Y_1 . Only when the real part of Y_1 has large negative values will a nonzero V_s be needed to keep the real part of the overall plume simulator impedance Z_s (= $Z_1 + Z_2$) nonnegative.

One therefore concludes that the MX missile plume simulator need only consist of a 1-port passive network parametrized by the input admittance Y_s (= $1/Z_s$):

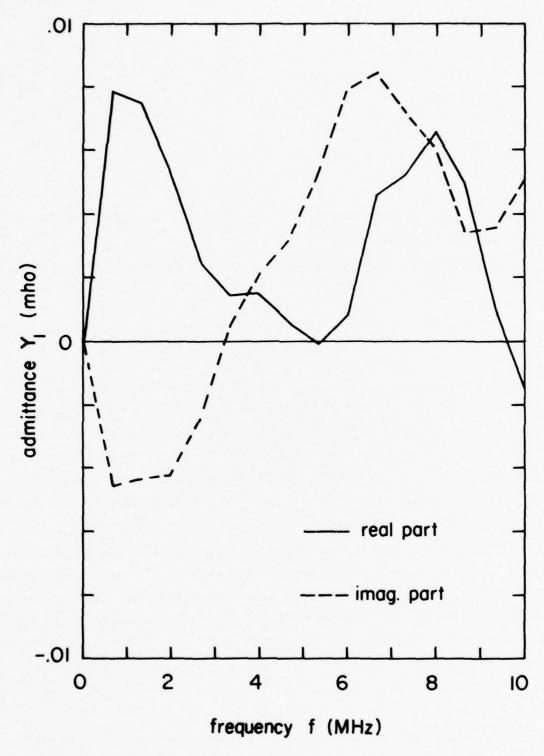


Fig. 8. Admittance Y_1 as a function of frequency f.

$$Y_s = Y_1 = \frac{Y_{in}}{I_{sc}/I_{noz} - 1}$$
 (7)

The values of this admittance are given in Fig. 8. The voltage source V_s in Fig. 7 can be chosen 0. This is because the open-circuit voltage V_{oc} due to the EMP simulator field is by itself strong enough to generate the nozzle current I_{noz} in the closed circuit in Fig. 7.

III. SYNTHESIS OF THE PLUME SIMULATOR

One proceeds to synthesize the plume simulator input admittance Y_s defined in Fig. 8, using only lumped LRC elements. The synthesis of the real part is first performed, followed by that of the imaginary part.

Synthesis of Re Y

The real part of Y_s in Fig. 8 shows two peaks. It can be realized with the lumped network circuit in Fig. 9, consisting of two resistors R_1 and R_2 , two inductors L_1 and L_2 , and one capacitor C_2 . For this network one has

Re
$$Y_s = \frac{R_1}{R_1^2 + (\omega L_1)^2} + \frac{R_2}{R_2^2 + (\omega L_2 - \frac{1}{\omega C_2})^2}$$
 (8)

with

$$\omega = 2\pi f \tag{9}$$

The values of the network elements are adjusted to fit the locations, widths and heights of the peaks. One finds that

$$R_1 = 100 \Omega$$
 $L_1 = 8 \mu H$ (10)

and that

$$R_2 = 167 \Omega$$
 $L_2 = 13.3 \mu H$ $C_2 = 30 pF$ (11)

The actual and simulated values of Re Y_s are plotted in Fig. 10 for comparison. One notices a disagreement at very low frequencies. As the frequency f approaches 0, the actual value of Re Y_s approaches 0 while the simulated value approaches a finite limit. This discrepancy, however, is not important, since both the currents I_{noz} and I_{sc} vanish at zero frequency. One can, if one chooses, bring the simulated value of Re Y_s down to 0 at zero frequency by adding a very large capacitor in series with I_1 and I_2 .

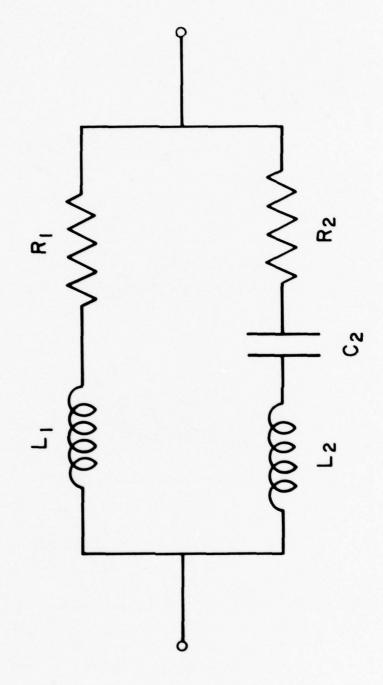


Fig. 9. Network for realizing the real part of the input admittance $\gamma_{\rm s}$ of the plume simulator.

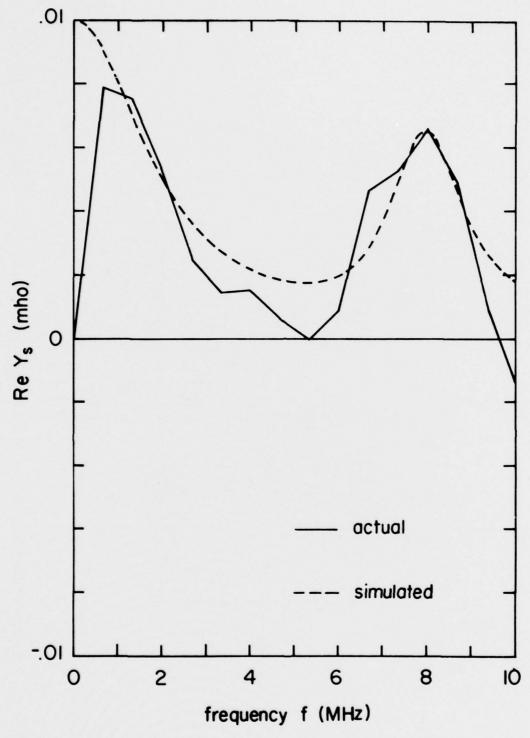


Fig. 10. Comparison between the actual and simulated values of $\mbox{ Re } \mbox{ Y}_{\mbox{\scriptsize S}}$.

Synthesis of Im Y

The imaginary part of the input admittance calculated from the network in Fig. 9 does not fit the actual values of $\operatorname{Im} Y_{S}$ specified in Fig. 8. Nevertheless, it is found possible to synthesize this difference by simply adding to the original network an inductor L_{3} and a capacitor C_{3} in parallel, as shown in Fig. 11. One then has

Im
$$Y_s = \frac{-\omega L_1}{R_1^2 + (\omega L_1)^2} + \frac{-(\omega L_2 - \frac{1}{\omega C_2})}{R_2^2 + (\omega L_2 - \frac{1}{\omega C_2})^2} + \omega C_3 - \frac{1}{\omega L_3}$$
 (12)

The added elements do not affect the real part of Y_s . By choosing the values

$$L_3 = 53 \text{ uH}$$
 $C_3 = 187 \text{ pF}$ (13)

one obtains a good fit between the actual and simulated values of ${\rm Im}\ {\rm Y}_{\rm S}$ in the most important frequency range between 1 and 3 MHz where the largest current components are found. The results are shown in Fig. 12. Again, the discrepancy at very low frequencies is unimportant.

Simulated Nozzle Current

The express purpose of the plume simulator is to deliver the correct nozzle current I_{noz} to the missile nozzle during EMP simulator testing. With the plume simulator made up of the passive network shown in Fig. 11, the simulated nozzle current is given by

$$I_{\text{noz}} = \frac{Y_{\text{s}}}{Y_{\text{in}} + Y_{\text{s}}} I_{\text{sc}}$$
 (14)

where Y_{in} is defined by Fig. 5 and I_{sc} by Fig. 6. The plume simulator input admittance Y_{s} is now given by

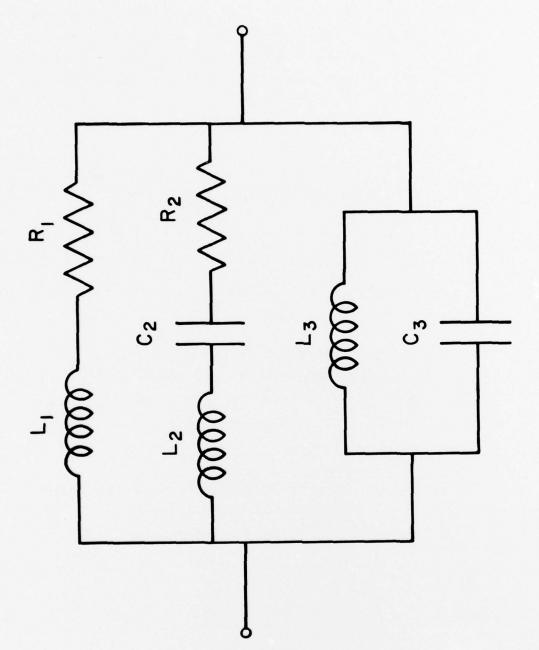


Fig. 11. Network for realizing the imaginary part of the input admittance $\gamma_{\rm S}$ of the plume simulator.

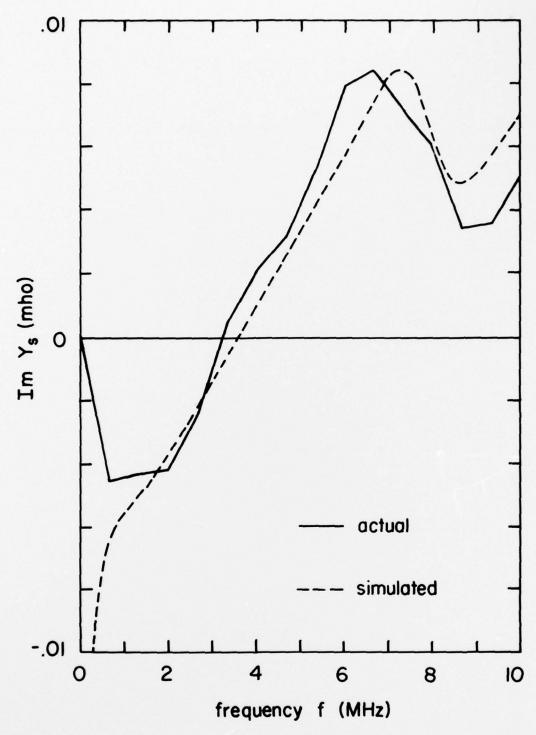


Fig. 12. Comparison between the actual and simulated values of $% \left(1\right) =\left(1\right) \left(1\right) =\left(1\right) \left(1\right) \left($

$$Y_{s} = \frac{1}{R_{1} + j\omega L_{1}} + \frac{1}{R_{2} + j\left(\omega L_{2} - \frac{1}{\omega C_{2}}\right)} + j\left(\omega C_{3} - \frac{1}{\omega L_{3}}\right)$$
 (15)

The actual and simulated nozzle currents are plotted in Figs. 13 and 14 for comparison. The agreement is good.

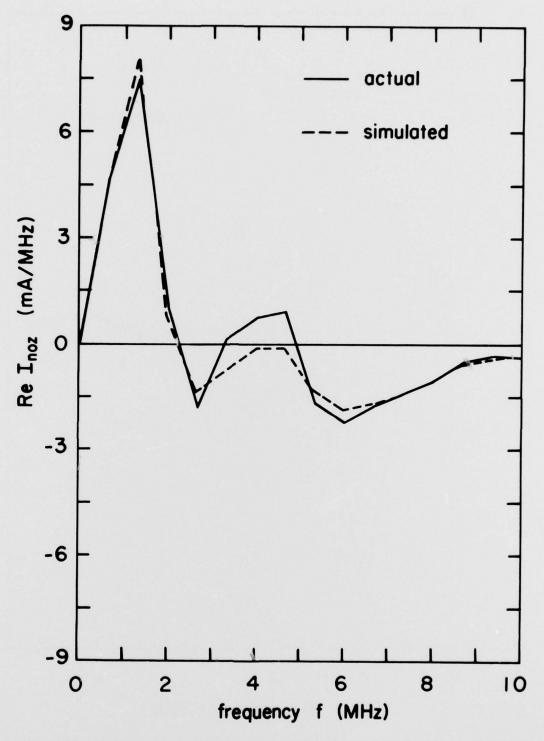


Fig. 13. Comparison between the actual and simulated values of the nozzle current : real part.

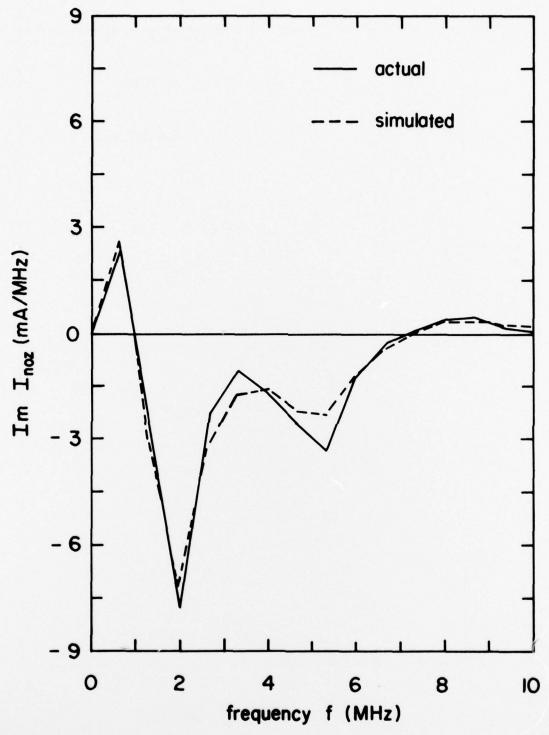


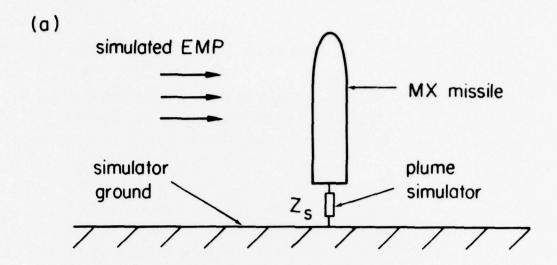
Fig. 14. Comparison between the actual and simulated values of the nozzle current: imaginary part.

IV. CONCLUSIONS

The interaction of an EMP with the exhaust plume of an MX missile and its effects on the missile itself can be simulated during EMP simulator testing by using a plume simulator. The plume simulator is to be connected between the nozzle of the missile and the simulator ground, as shown in Fig. 15a. At EMP frequencies the exact geometry of the connection is immaterial as long as the electrical contact is good. The structure of the plume simulator for a given theoretical model of the plume [1] is determined and shown in Fig. 15b. It consists of a passive lumped network characterized by an input impedance $Z_{\rm g}$.

For an incident EMP with an electric field intensity of 1 V/m, the peak value of the current flowing through the plume simulator is estimated to be of order 50 mA and the peak value of the voltage across its terminals of order 10 V. Typically the electric field intensity expected in an EMT is of order 50 kV/m. Therefore the plume simulator should be constructed to withstand a current of order 2500 A and a voltage of order 500,000 V.

The advantages of a plume simulator built entirely out of passive elements are many and easy to grasp. First, there is no need to fabricate and service current or voltage generators such as enormous charged capacitors. Second, the delicate technical problem of synchronizing the switching-on of these current or voltage generators with the launching of the EMP simulator field on the missile is thereby eliminated. Third, the same passive plume simulator can be used for different field-strength settings of the EMP simulator.



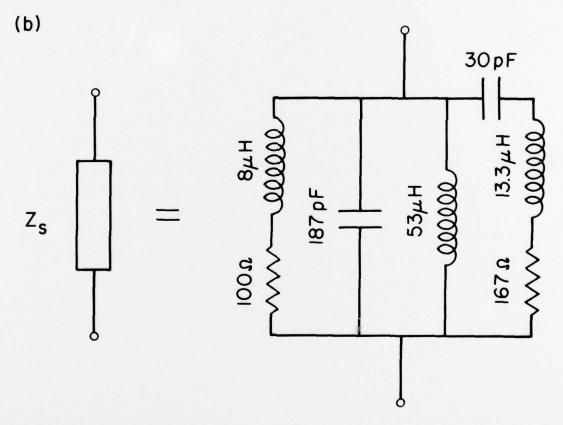


Fig. 15. Structure of a plume simulator for the MX missile.

REFERENCE

[1] Chang, S.K., F.M. Tesche, D.V. Giri, "EMP Coupling to an In-Flight MX Missile in the 0 - 20 Km Altitude Regime", Science Applications, Inc., Berkeley, CA, final report on Contract DNA001-77-C-0184, March 1978.

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